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Experimental design and numerical investigation of a photoacoustic sensor for a low-power, continuous-wave, laser-based frequency-domain photoacoustic microscopy

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Abstract. We have developed a photoacoustic (PA) sensor using a low-power, continuous-wave laser and a kHz-range microphone. The sensor is simple, flexible, cost-effective, and compatible with commercial optical microscopes. The sensor enables noncontact PA measurements through air, whereas most current existing PA techniques require an acoustic coupling liquid for detection. The PA sensor has three main components: one is the chamber that holds the sample, the second is a resonator column used to amplify the weak PA signals generated within the sample chamber, and the third is a microphone at the end of the resonator column to detect the amplified signals. The chamber size was designed to be 8 mm × 3 mm as the thermal diffusion length and viscous-thermal damping of air at room pressure and temperature are 2 and 1 mm, respectively. We numerically and experimentally examined the effect of the resonator column size on the frequency response of the PA sensor. The quality factor decreased significantly when the sample chamber size was reduced from 4 mm × 3 mm to 2 mm × 3 mm due to thermos-viscous damping of the air. The quality factor decreased by 27%, demonstrating the need for optimal design for the sample chamber and resonator column size. The system exhibited noise equivalent molecular sensitivity (NEM) per unit bandwidth (NEM/ $\sqrt{\Delta f}$) of $\sim 19,966 \text{ Hz}^{-1/2}$ or $33 \times 10^{-21} \text{ mol}$ or 33 zeptomol, which is an improvement of 2.2 times compared to the previous system design. This PA sensor has the potential for noncontact high-resolution PA imaging of materials without the need for coupling fluids. © The Authors. Published by SPIE under a Creative Commons Attribution 4.0 Unported License. Distribution or reproduction of this work in whole or in part requires full attribution of the original publication, including its DOI. [DOI: 10.1117/1.JBO.24.12.121912]

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1 Introduction

The photoacoustic (PA) effect was first observed in the year 1880 by Alexander Graham Bell.¹ The PA effect was used in spectroscopy techniques throughout the mid-1900s, but the invention of the laser has significantly changed the paradigm of the PA technique because of the improved sensitivity provided by pulsed laser excitation.^{2–13}

Photoacoustic microscopy (PAM) is a hybrid imaging modality that uses an optical technique for excitation and an acoustic technique for detection. The general principle is based on the generation of acoustic waves due to transient pressure variation caused by means of absorption of a short (typically nanosecond) light pulse. As high-resolution deep tissue imaging is possible using acoustic resolution PA imaging,^{14–17} PA imaging finds several potential applications providing anatomical,¹⁸ functional,¹⁴ molecular,¹⁴ flow dynamic,¹⁵ and metabolic contrast *in vivo*.¹⁵

Microscopy applications can use either acoustic resolution PAM (AR-PAM) or optical-resolution PAM (OR-PAM), or a combination of both, depending on the application and imaging target.¹⁹ The lateral resolution of AR-PAM depends on the acoustic focus of the transducer which is of the order of tens

of micrometers with a maximum imaging depth of several millimeters whereas the lateral resolution of OR-PAM depends on the diffraction limit of the optical lens with a maximum possible imaging depth of about 1.2 mm.^{20,21} Hence, OR-PAM has been used to image biological samples and small animals with various endogenous or exogenous contrasts²⁰ whereas AR-PAM finds both preclinical and clinical applications as deep imaging of biological tissues is possible at lower ultrasound frequencies.^{22,23}

Researchers are also interested in noncontact PA systems for remote sensing. Most of the noncontact PA systems are based on an interferometric principle.^{24–28} However, interferometric techniques can be complex, bulky, expensive, and difficult to integrate with the existing optical microscopes.^{24,29} A Fabry–Perot sensor is another option, but it is still a contact technique as the Fabry–Perot sensor has to be coupled to the sample.^{10,30} Researchers have investigated using a fiber sensor, which also required expensive focusing and collection optics. Optical beam deflection techniques are another simple method for PA wave detection. However, this approach requires two lasers for pumping and probing and a large footprint area for the set up because the probe beam needs to travel a relatively long distance (for increasing deflection angle) before detection to improve sensitivity.^{31,32} There are also other air-coupled transducers available, but air-coupled transducers are unfocused, bulky, and exhibit poor SNR, which makes imaging more complicated.³³

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Conventional PA imaging systems require expensive ultrafast lasers with nanosecond pulse widths, an ultrasound transducer (or optical-based ultrasound detection), and an aqueous coupling medium for the PA waves to propagate through.^{10,17,19} Wide clinical adoption has been hindered due to its expense, technical complexity, and cumbersome size. Considerable efforts have been made to use low-cost continuous-wave (CW) lasers instead of ultrafast laser for PA imaging since 1990.^{34–36} Petschke and La Rivière³⁵ theoretically investigated the possibility of using chirped CW diode lasers, which would exhibit 20 to 30 dB lower SNR than the typical pulsed laser-based systems, but its compactness and relatively low cost could potentially outweigh the lower SNR in select applications.^{34,35,37}

Recently, we performed noncontact PA studies using a low-power CW laser.³⁸ The weak PA signal generated by the low-power CW laser was amplified using an acoustic resonator and measured using a kHz microphone. The system was studied on a red blood cell (RBC) and found exhibited a noise-equivalent detection sensitivity of 43,887 hemoglobin molecules (72.88×10^{-21} mol or 72.88 zeptomole). The minimum pressure detectable limit (MPDL) of the sensor was 19.1 μ Pa. The technique was inspired by low-power CW-based PA spectroscopy as it exhibited trace gas detection sensitivity of parts per billion.^{3–5,11–13,39–47}

The magnitude of the PA signal is proportional to the thermal diffusion length (μ)/thermal diffusivity of the sample which is a function of optical and thermal properties of the material. As the PA signal is generated by intensity-modulated light, the thermal diffusion length (μ) would become a function of the modulation frequency of the light and is given by

$$\mu = \frac{k}{\pi \rho f c}, \quad (1)$$

where k is the thermal conductivity, ρ is the density, c is the specific heat, and f is the modulation frequency. Figure 2(a) shows a schematic diagram of the most commonly used PA cell for spectroscopy studies. The sample is enclosed in an airtight metal chamber and excited over the surface. The PA signal is due to pressure or stress generated in the gas adjacent to the sample. This PA cell configuration may not be suitable for high-resolution optical resolution PAM because exciting the sample using a high-magnification 100 \times objective is difficult due to its shorter working distance. The typical working distance of a 100 \times objective is around 0.17 mm but the sample in the PA cell should be positioned at least 1 mm away from the optical window to avoid acoustic damping caused by the walls of the PA cell.

The PA sensor for high-resolution OR-PAM proposed in this work has a different configuration compared to those used for PA spectroscopy. Figure 2(b) shows the schematic diagram of the PA sensor. The sensor works in a transillumination configuration. The sample is excited at the sample–glass interface. The thermal properties of the glass substrate play a vital role in the PA generation as it might lead to heat conduction outside the chamber, which would result in a decrease in the PA signal. The influence of the glass on the PA generation due to the back and surface excitation of the sample (Fig. 2) was investigated. The thermal diffusion length and viscous-thermal damping of the air will also play important roles in the loss of PA signal depending on the size of the PA cell. The effects of these

factors on the PA signal are numerically and experimentally investigated.

One advantage of the present PA sensor would be that it could be used in both epi- and trans-illumination microscopes. The earlier ones were bulky and sturdy because most of the sensors were made of metals. The developed sensor is thin, compact, inexpensive, and flexible. The system can be easily integrated/attached into any systems or object to study the optical properties. In addition, the configuration of the sensor would enable multimodal imaging such as optical, fluorescent, Raman, and PA imaging of samples. As the PA spectroscopy technique can be used to study samples in any state of matter (solid, liquid, and gas), the PA sensor is also expected to find similar applications in studying both solid and liquid samples. This study investigates the use of the PA sensor in OR-PAM applications, with a focus on cell design using numerical simulation techniques and experimental verification.

2 Materials and Methods

2.1 Procurement of Blood

Blood was collected by netCAD (Vancouver, Canada), the research division of Canadian Blood Services, under protocol 2013-001, which involves standard Canadian Blood Services collection and testing procedures of whole blood and delivery overnight at 4°C, with continuous monitoring during shipment to ensure no temperature deviations occur. This procedure has been approved by the research ethics boards.

2.2 Blood Sample Preparation

The guidelines on handling the blood were followed in accordance with the recommendations of the International Society for Clinical Hemorheology and the European Society for Clinical Hemorheology and Microcirculation. The blood was centrifuged at room temperature at $2000 \times g$ for 6 min to separate the plasma and their Buffy coat. Isotonic phosphate-buffered saline (PBS) was used to wash RBCs two times. The centrifuged RBCs were then dispersed in PBS for the present studies. RBCs were smeared on microscope slides for studies.

2.3 Experimental Setup

A simple dual-modal optical/PAM system was developed to characterize the developed PA sensor (Fig. 1). The PA system is composed of an optical imaging system, a laser for PA excitation, a motorized translation stage to raster scan the sample, and the PA sensor. The optical imaging system was developed in inverted microscope configuration. The system contains an infinity-corrected 40 \times and 0.25 NA Olympus objective (O), and a Ximea xiQ camera (Ximea). This configuration of this system was simpler than our earlier reported optical microscope system.³⁸ The objective was moved to excite the sample in the earlier configuration, which required a complex optical configuration to keep the laser beam coaxial with the optic axis of the objective. The present configuration uses a fixed optical set-up in which the light beam is coupled directly to the stationary objective for excitation and the sample is moved.

A helium–neon laser (Spectra-Physics) with a beam diameter of 1 mm and a green laser (CrystaLaser) with a beam diameter 0.36 mm were the optical source of the PA system. The power of the laser was 4 mW. The solid green line in Fig. 1 shows the laser path in the PA microscope system which was focused

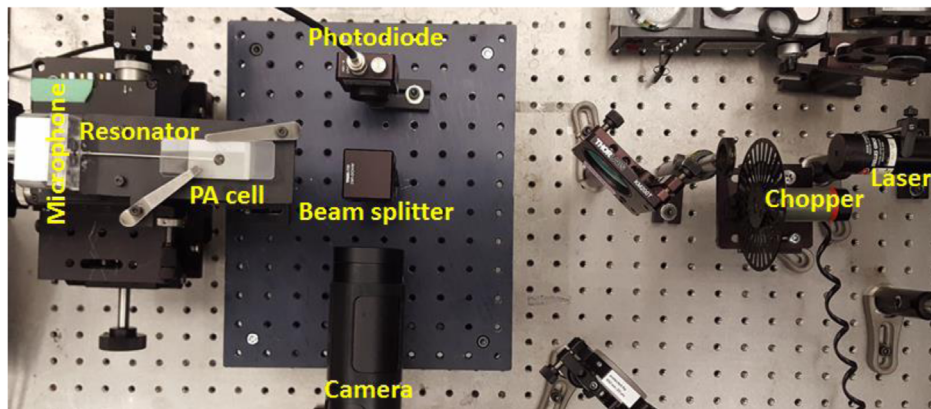
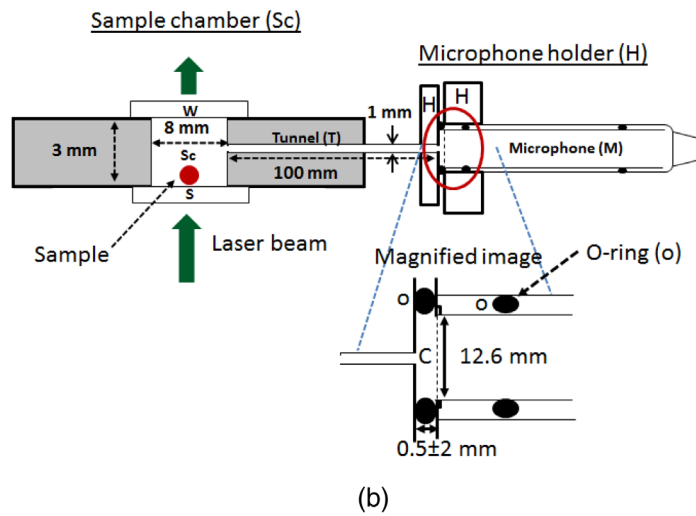
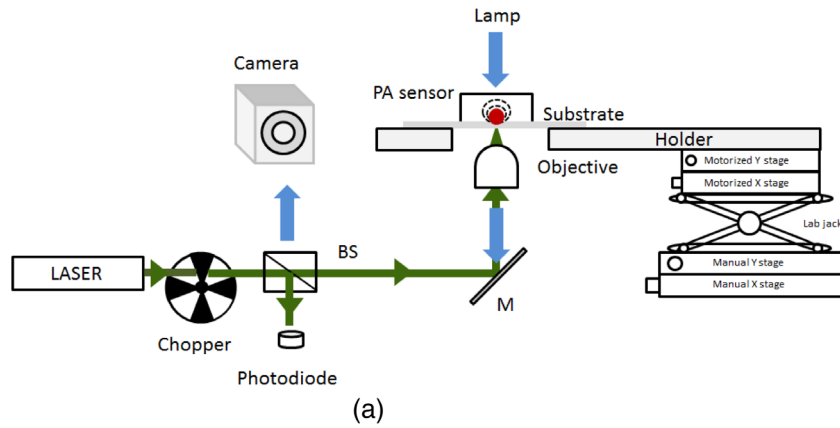


Fig. 1 A schematic diagram of the optical/PAM system. The solid green line and the blue arrows represent the laser beam path and polychromatic light beam direction in the PA and the optical microscope system, respectively. (b) The schematic diagram of PA sensor where Sc is the sample chamber, S is the glass substrate, W is the glass window, T is the tunnel, C is the coupler, O is the O-ring, H is the holder, and M is the microphone. (c) The photograph of a top view of the system.

on the sample by an objective. A cube beam splitter was used to sample a small portion the input beam toward the photodiode (PD) for reference, and also used to direct the trans-illuminated LED light, which counterpropagates in the direction of the laser beam, from the sample to the camera positioned opposite to PD. The laser was modulated using a mechanical chopper (SR540,

Stanford Research Systems) to generate a pulsed output. The PA signal was measured using a lock-in amplifier (SR850, Stanford Research Systems). The system was automated using custom Matlab software (Mathworks).

The translation system consisted of manual X-Y-Z stages (Thorlabs) for interrogating the sample and motorized linear

X-Y scanning stages from Zaber Technologies, for raster scanning the sample for PA imaging.

The measurement is made using a lock-in amplifier (SR850). The lock-in amplifier contains two main components: the pre-amplifier and the phase-sensitive detector (PSD). The input of the lock-in amplifier was first amplified before entering a PSD. The PSD detects the signal with respect to a reference by multiplication of the amplified PA signal with the lock-in reference signal produced by the crystal oscillator of lock-in amplifier in accordance with an input reference signal (in our case, the input reference signal is from the chopper). The output of the PSD is the product of two sine waves:

$$\begin{aligned} V_{\text{psd}} &= V_{\text{sig}} V_L \sin(\omega_r t + \theta_{\text{sig}}) \sin(\omega_L t + \theta_{\text{ref}}) \\ &= \frac{1}{2} V_{\text{sig}} V_L \cos([\omega_r - \omega_L]t + \theta_{\text{sig}} - \theta_{\text{ref}}) \\ &\quad - \frac{1}{2} V_{\text{sig}} V_L \cos([\omega_r + \omega_L]t + \theta_{\text{sig}} + \theta_{\text{ref}}), \end{aligned} \quad (2)$$

where V_{sig} is the signal amplitude, V_L is the reference amplitude, ω_s is the signal frequency, ω_r is the reference frequency, θ_{sig} is the signal phase, and θ_{ref} is the reference phase.

The SR850 is a dual-phase lock-in amplifier with a second PSD measure the signal at 90 deg out of phase with respect to the first PSD. The output is a contribution from both PSDs. The magnitude of the signal is given as

$$R = (V_{\text{psd1}}^2 + V_{\text{psd2}}^2)^{1/2} = V_{\text{sig}}. \quad (3)$$

In this work, first, the SNR of the PA system was investigated. The SNR mainly depends on the input configuration of the lock-in amplifier and the power of the laser. The system used low-power lasers. The lock-in amplifier response for the low-power laser was studied. Carbon black was used as the sample. The noise of the lock-in amplifier depends on filter bandwidth which is a function of signal integration time constant. The SNR of the system for an integration time of 10 ms, 300 ms, 1 s, and 3 s was studied (please see the [Supplemental Materials](#)). A lock-in amplifier functional at a low pass filter setting/single RC filter with a frequency response of 6 dB/oct and a time constant 300 ms exhibited a low SNR of 3.31×10^{-6} (please refer to the [Supplemental Materials](#)).

2.4 Photoacoustic Sensor

The PA sensor for high-resolution OR-PAM is shown in Fig. 2(b). The sample is excited at the sample-glass interface. The effect of glass on PA generation is studied experimentally. Two optical configurations (Fig. 2) are used for these studies. The sample is excited in front and backside in these configurations. The PA study using both configurations showed that the front surface excitation [Fig. 2(a)] exhibited about 74% greater signal compared to the backside excitation [Fig. 2(b)] (please see the [Supplemental Materials](#)).

A schematic of the PA sensor is shown in Fig. 1(b). The PA sensor is made up of acrylic photopolymers and aluminum. The PA sensor has two main compartments, one for holding the sample (called sample chamber, Sc) and the other a resonant column coupled to the microphone (M). The Sc is a hollow cylindrical of diameter 8 mm and a height 3 mm. The sample chamber is built in acrylic photopolymers using a three-dimensional printer (ProJet 3510 HDPlus 3D printer). The top of the sample

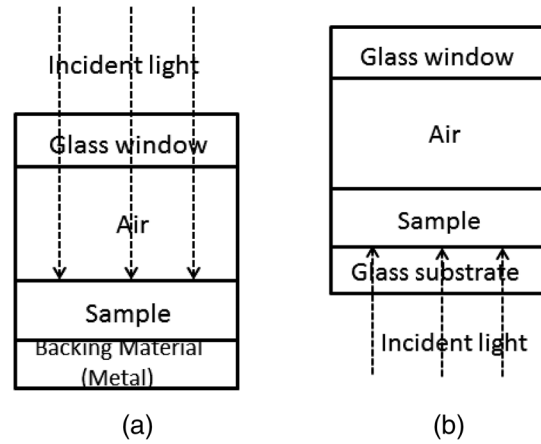


Fig. 2 A schematic diagram of (a) a typical PA cell configuration used for spectroscopy studies and (b) the PA cell used for the high-resolution PAM in these studies.

chamber is closed by an optical window (microscope slide), whereas the bottom of the chamber contains the sample (S) for study. A small resonant column (T) exists between the sample chamber and the microphone (M) which is a cylinder of diameter 1 mm and varying length. A cylindrical coupler (C) of inner diameter 12.6 mm (equivalent to the diameter of the microphone) and height 0.5 mm is used to join the resonator column to the microphone because of a mismatch of their size (1 and 12.7 mm, respectively). The microphone is housed in a specially designed rectangular holder (H) with four holes at the corners for screws. The resonator, the coupler, and the microphone holder are made of aluminum. The microphone holder H is screwed airtight with respect to the coupler (C) using a rubber O-ring of inner diameter 16 mm and thickness 2 mm between them. The gap between the resonator and the microphone surface may vary between 0.5 ± 0.2 mm due to the squeezing of rubber O-ring.

The resonator behaves like a resonance tube with an acoustic node formed at a closed-end (microphone end) and antinode formed at an open-end (sample chamber end). Resonance is observed when a standing acoustic mode is formed in the resonator. As the frequency of the acoustic mode is inversely proportional to the length of the resonator, the effect of the resonator length on the PA signal is studied. Resonator column lengths of 100, 200, and 400 mm were chosen for the study. The length was limited to 400 mm because of the size of the translation stage of most microscopes lies in this range.

The effect of the Sc size on the PA generation was also investigated. The sizes of the Sc chosen for these studies were 2 mm \times 3 mm, 4 mm \times 3 mm, 6 mm \times 3 mm, and 8 mm \times 3 mm. Sizes below 2 mm were not considered because of acoustic damping (the viscous thermal damping distance of air is ~ 2 mm at KHz frequencies). A half-inch free-field microphone of type 4189 with a Type 2699 preamplifier (Brüel & Kjær Sound & Vibration Measurement, Pointe-Claire, Quebec, Canada) was used to measure the PA signal. The microphone has an almost flat frequency response from 50 Hz to 20 kHz with a sensitivity of 50 mV/Pa.

2.5 Characterizing of a Photoacoustic Sensor

The frequency response of the PA sensor was measured by modulating the output of the helium-neon laser using chopper

from frequency 50 to 1100 Hz. The frequency of the chopper was varied by changing its driver voltage by means of the potentiometer available in the system.⁴⁸ The lock-in amplifier was programmed at low-pass filter setting with a frequency response of ± 6 dB/oct and time constant 300 ms. The integration time can be increased at the cost of imaging speed to improve the SNR if the laser power is low.

3 Photoacoustic Theory of a Cylindrical PA Sensor

The sound field distribution in the cylindrical column representing the PA chamber is described by the Helmholtz equation⁴⁹

$$\left[\frac{1}{r} \frac{\partial}{\partial r} \left(r \frac{\partial}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2}{\partial \varphi^2} + \frac{\partial^2}{\partial z^2} \right] p(r, \varphi, z, \omega) + k^2 p(r, \varphi, z, \omega) = i\omega \frac{(\gamma - 1)}{V_C} H(r, \varphi, z, \omega), \quad (4)$$

where V_C is the volume of the PA chamber, γ is the ratio of the specific heats, k is the wavenumber, and $H(r, \varphi, z, \omega)$ represents the Fourier transform of the heat source term $H(r, \varphi, z, t)$. Here, the heat source represents the sample heated by the laser. The sample with an absorption coefficient α and excited with light of intensity $I(r, \varphi, z, \omega)$ exhibits heating given by $H(r, \varphi, z, \omega) = \alpha I(r, \varphi, z, \omega)$, where I is the amplitude of the laser intensity.

Solving the above inhomogeneous wave, Eq. (6) gives the acoustic modes supported by the cylindrical column⁵⁰

$$p(r, \varphi, z, \omega) = \sum_j A_j(\omega) p_j(r, \varphi, z), \quad (5)$$

where $A_j(\omega)$ is the amplitude of the acoustic mode $p_j(r, \varphi, z)$ and ω_j ($\omega_j = ck_j$ or $2\pi f_j = ck_j$) is the eigenfrequency. The equation is solved by applying a sound hard wall boundary condition the normal derivative of the pressure ($\frac{\partial p}{\partial n} = 0$). The solution gives the supported eigenmodes of the PA sensor. The amplitude $A_j(\omega)$ and the frequency f_{lmn} of the eigenmodes supported by the cylindrical PA sensor is given as

$$A_j(\omega) = i \frac{A_j \omega}{\omega^2 - \omega_j^2 + i\omega \omega_j / Q_j}, \quad (6)$$

where

$$A_j(\omega) = \frac{\alpha(\gamma - 1)}{V_C} \int p_j^* \cdot IdV, \quad (7)$$

and

$$f_{lmn} = \frac{c}{2} \sqrt{\left(\frac{l}{L}\right)^2 + \left(\frac{\alpha_{mn}}{r_c}\right)^2}, \quad (8)$$

where α_{mn} is the zeros of the derivative of the Bessel functions, L is the length of the PA cell, r_c is the radius of the PA cell, l is the longitudinal quantum number, m is the azimuthal quantum number, and n is the radial quantum number.^{49,50}

3.1 Photoacoustic Cell Simulations

The acoustic properties of the PA chamber were studied using finite element software from COMSOL (Burlington, Massachusetts). The pressure acoustics, frequency domain (acpr) user interface was used for the present study. The governing equations in the pressure acoustic model were formulated using the total pressure in a scattered field formulation. The solution gives the acoustic variation/excess pressure (p) to the ambient pressure. The solved Helmholtz wave equations are⁵¹

$$\nabla \cdot \frac{1}{\rho_c} (\nabla \rho_t - q_d) - \frac{k_{eq}^2 \rho_t}{\rho_c} = Q_m,$$

where

$$\rho_t = \rho_s + \rho_b, \quad (9)$$

$$k_{eq}^2 = \left(\frac{\omega}{c_c}\right)^2,$$

$$c_c = c \quad \text{and} \quad \rho_c = \rho,$$

where c is the speed of sound, ρ is the density of material, t is the time, ρ_t is the total acoustic pressure, ρ_s is the scattered pressure field, ρ_b is the background pressure field, k_{eq} is the amplitude of the wave vector k , ω is the angular frequency, and q_d and Q_m are acoustic dipole and monopole source terms, respectively. The subscript c to the density ρ and the speed of sound c represent the complex speed of sound and density, respectively.³⁶ Lossy media such as porous materials, highly viscous fluids, etc., are modeled by using the complex-valued speed of sound (c_c) and density (ρ_c). Since air is an acoustic transport medium, c_c and ρ_c in our case become c and ρ , respectively.

The purpose of this study is to optimize the PA cell for maximum output. The generation of the PA signal amplitude is inversely proportional to the cell volume. Larger cell volumes result in weaker signals due to the damping within the cell. Conversely, for small cell volumes, the thermal diffusion length and viscous-thermal damping distance of air progressively become dominant. The thermal diffusion length and viscous-thermal damping distance of air were also taken into account in designing the cell. The thermal diffusion length and viscous-thermal damping of gas are dependent on frequency, temperature, and pressure. At room temperature, pressure, and at a frequency of 100 Hz, the air has a thermal diffusion length and viscous-thermal damping distance of 0.2 and 0.1 mm, respectively.¹¹ A PA sensor to work in a broad range of modulation frequencies should need minimum dimensions of 1 to 2 mm to avoid any PA signal saturation due to acoustic damping of PA cell walls.¹¹ The influence of this effect on the PA signal was investigated by adding thermoacoustic physics to the pressure acoustic model.

3.1.1 Study of the effect of sample chamber on PA signal

First, the effect of Sc size on the PA signal was investigated. The diameters of the Sc chosen for the studies varied between 2 and 8 mm. The height of the chamber was 3 mm. The height was chosen to 3 mm in order to leave enough room to accommodate the sample, and 1 mm resonator though 2 mm enough to avoid the thermo-viscous losses caused by cell surface. The length of the resonator column used in the study was 100 mm.

The PA source is a sample heated by a laser, which is represented by a heat source (Q) in the simulation. The heat source was simulated to exhibit a Gaussian heat distribution (a TEM₀₀ mode laser was used for the ORPAM). The Gaussian heat distribution is given as

$$Q_0 = Q \exp\left(\frac{-r^2}{2\sigma_0^2}\right), \quad (10)$$

where σ_0 is the laser beam Gaussian spread function and Q is the heat input amplitude. Since the laser beam was focused using 10×, 0.25 NA objective, the irradiance produced was 1 W/mm². In experiments, carbon black was used. Carbon black has a heat conversion efficiency from 0.5 to 1 depending on the material. Assuming the light to the heat conversion efficiency of the sample to be 0.5, the heat produced by 1 W/mm² irradiance would be 1 J/mm².

When the sound waves propagate in a fluid bounded by walls, viscous and thermal boundary layers will be created close to the PA cell surfaces due to the interaction of the fluid with solid PA cell wall and the difference in the thermal conductivities of the liquid and the solid wall. In the case of a small PA chamber viscous and thermal losses can be significant. A no-slip ($u = 0$) and isothermal [$-\vec{n} \cdot (-k \cdot \nabla T) = 0$] conditions were employed to all the boundaries of the simulation region. Nodal constrain boundary condition was chosen in order to account any attenuation caused by sharp edges of the PA chamber.

Incorporating the microphone and lock-in response in the numerical simulation. The frequency response of the PA sensor is the convolution of the response of the microphone and a lock-in amplifier. The microphone response at low frequencies is not flat. The microphone frequency response can be described by simpler acoustic diaphragm impedance models.⁵² Figure 3 shows the equivalent circuit of the acoustic diaphragm impedance of the microphone. The components represent stiffnesses, masses, and damping of the electromechanical system. The total impedance of the circuit is

$$z = \frac{1}{i\omega C_{ac}} + R_{ac} + i\omega L_{ac}, \quad (11)$$

where C_{ac} is the diaphragm system compliance, L_{ac} is the diaphragm system mass, and R_{ac} is the diaphragm system resistance.

The effect of the microphone is accounted for in the simulation by applying serial coupling RCL impedance boundary condition with values of C_{ac} , L_{ac} , and R_{ac} at the microphone end (Table 1).

The microphone response L_0 (for a constant volume displacement source):⁵³

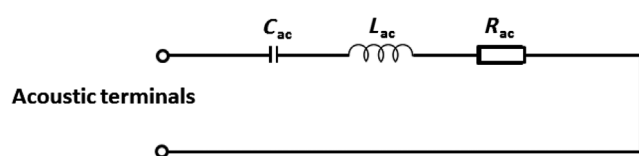


Fig. 3 A simplified model representing the acoustical impedance of the microphone.

Table 1 Values of the simplified microphone model shown in Fig. 3.

Acoustic parameter	Equivalent electric value
R_{ac}	2.83×10^{-13} Ns/m ⁵
C_{ac}	896 m ⁵ /N
L_{ac}	56.2×10^6 kg/m ⁴

$$L_0 = 10 \log\left(\frac{\langle P_{mic} \rangle^2}{P_{ref}^2}\right), \quad (12)$$

where $\langle P_{mic} \rangle^2$ is the root mean square pressure at the measurement microphone and P_{ref} is the reference pressure.

Since the output of the microphone is preamplified by the amplifier in the microphone, the frequency response of the system will be the combination of both the electrical response of the preamplifier and the acousto-mechanical response of the microphone.⁵² The electrical circuit of the combined system that comprising of both the microphone and the preamplifier is shown in Fig. 4.

The electrical frequency response of the circuit is given as

$$\frac{V_0}{V_{oc}} = \frac{C_m}{C_m + C_i} \cdot \frac{j\omega(C_m + C_i)R_i}{1 + j\omega(C_m + C_i)R_i} \cdot g \cdot \frac{1}{1 + jC_c R_0}, \quad (13)$$

where V_{oc} is the open circuit microphone voltage, V_0 is the output voltage, C_m is the microphone capacitance, C_i is the preamplifier input capacitance, R_i is the preamplifier input resistance, g is the amplifier gain, R_0 is the preamplifier output resistance, and C_c is the cable capacitance whose typical values are given in Table 2.

The output of the PSD of the lock-in amplifier contains many signals. The main components are the sum and difference frequency of the input and the reference signals. The low pass filter at the PSD output filters out all the unwanted AC signals that include 2F (sum frequency) and the noise components but records the dc component which is the input signal detected by a lock-in at the reference frequency. As the measurement was made using the lock-in amplifier set at a first-order low pass filter configuration with a time constant (τ), the transfer function of the lock-in amplifier, in this case, is $T(\omega) = \left| \frac{V_o}{V_i} \right| = \frac{1}{\sqrt{1+(\omega\tau)^2}}$, where V_o is the output, V_i is the input, and ω is the angular

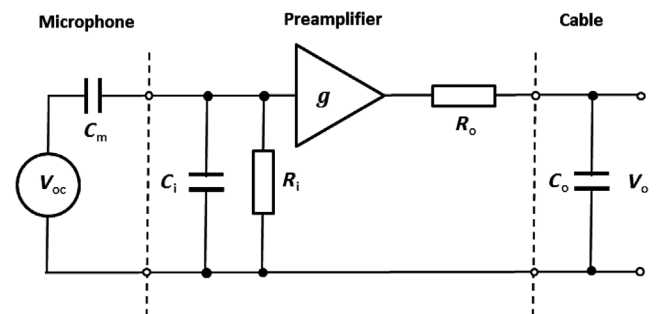


Fig. 4 An simple model for calculation of the electrical frequency response of a microphone and its preamplifier loaded with a cable.

Table 2 Values of electrical circuit elements of the simplified model shown in Fig. 4.

Symbol	Typical value
V_{oc}	1 μ V to 50 V
C_m	18 pF
C_i	0.2 pF
R_i	10 G Ω
g	0.995
R_0	30 Ω
C_c	3 nF

frequency. Hence, the signals of frequencies less than $1/\tau$ were passed but frequencies $\omega\tau \gg 1$ were attenuated by $T \rightarrow 1/\omega\tau$. The transfer function of the lock-in in this configuration was measured experimentally using a nanosecond response PD from Thorlabs (please refer to Fig. S3 in the Supplemental Materials). The electrical frequency response of the microphone [Eq. (13)] and the experimentally measured transfer function (Fig. S3 in the Supplemental Materials) of the lock-in amplifier were multiplied with the result of the numerical simulation to generate the PA response [Eq. (12)] of the cell. Finally, the PA response was squared according to Eq. (3) as the PSD of lock-in involved the multiplication of the input and the reference signal.

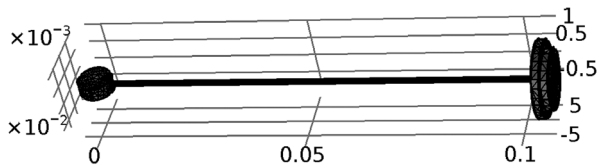
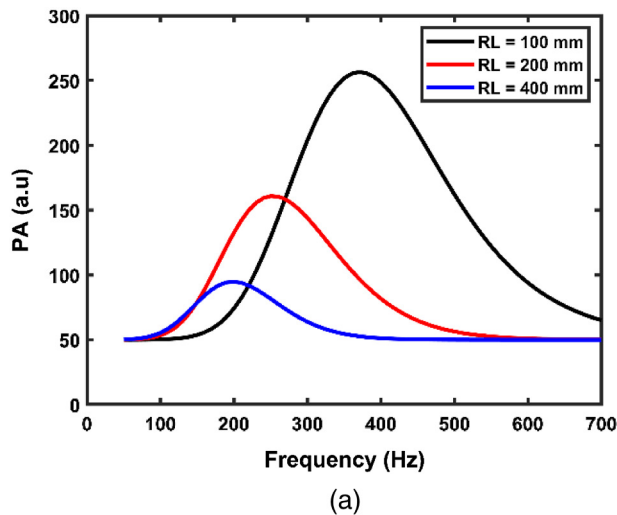


Fig. 5 The meshing of the PA cell of chamber size 8 mm \times 3 mm with a resonator of length 100 mm.



3.1.2 Study of the effect of resonance tunnel on PA signal

The resonance of the PA cell can be tuned by changing the length of the resonator column/Helmholtz resonator. The narrow cylindrical channel connecting Sc and microphone acts as the Helmholtz resonator. The Sc end of the resonator acts as an open-end while the microphone end acts as a close end. The resonator of lengths 100, 200, and 400 mm was investigated. The air was the acoustic transport medium. The resonator of diameter 1 mm is chosen for the study. The entire geometry was meshed using the extremely fine precondition in COMSOL. Figure 5 represents the 8 mm \times 3 mm PA sensor with a resonator of length 100 mm used in COMSOL. The sensor is meshed by extremely fine precondition for numerical simulation.

The configuration was solved for the pressure distribution with frequency ranging from 50 to 1000 Hz.

4 Results and Discussion

Figure 6 shows the simulated and measured PA signals as a function of frequency for resonator lengths 100, 200, and 400 mm. The resonance peak was decreased with an increase in resonator column length, which is expected, as the frequency is inversely proportional to the resonator length [Eq. (8)]. The experimental results exhibited a similar trend to the numerical simulations. Table 3 compares the resonance frequencies of the

Table 3 Comparison of resonance frequencies obtained from experimental measurements and numerical simulations for various resonator lengths.

Sample chamber size 8 mm \times 3 mm	Frequency (Hz) (numerical)	Frequency (Hz) (experiment)	Percentage difference
Resonator length $R_l = 100$ mm	370	377	1.87
$R_l = 200$ mm	240	231	3.82
$R_l = 400$ mm	200	174	13.9

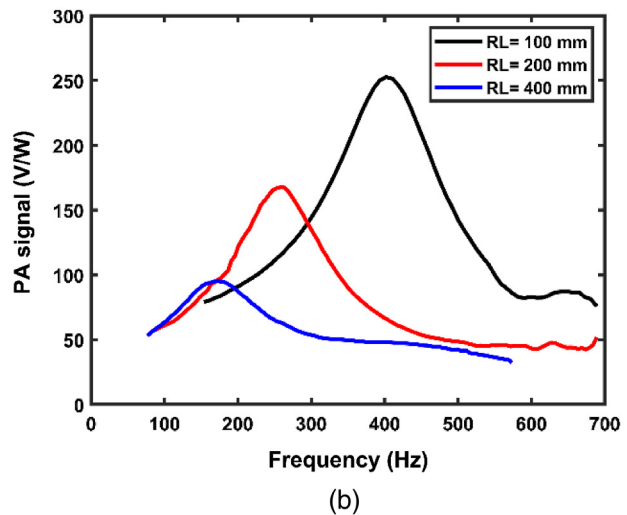


Fig. 6 The frequency response of the PA sensor for various resonator lengths from (a) theory and (b) experiment.

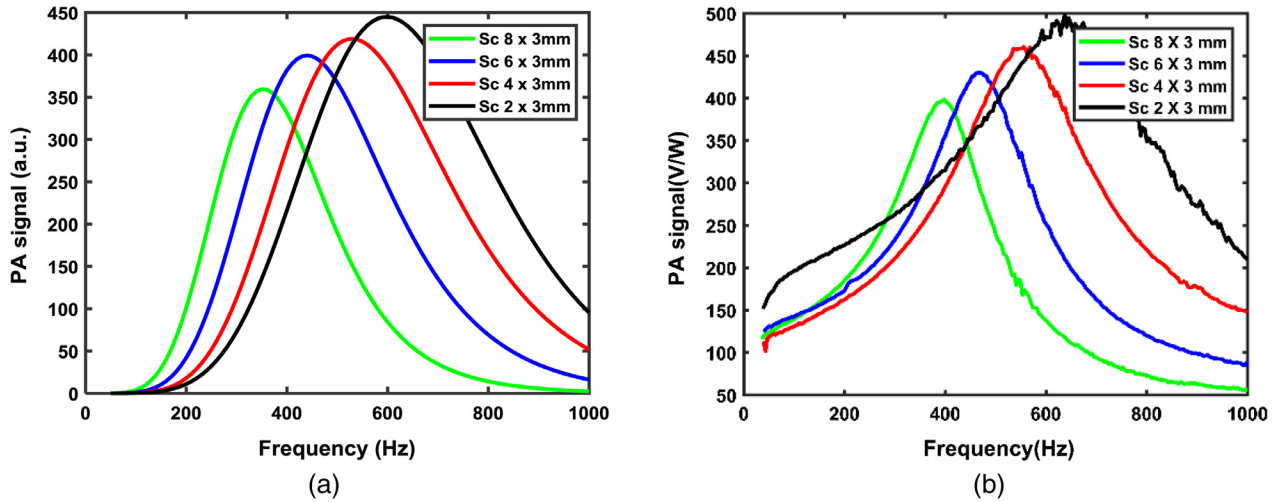


Fig. 7 The frequency response of the PA sensor for sample chambers of various sizes (a) simulation and (b) experiment.

sensor with various resonator lengths obtained from both numerical simulations and experimental measurements.

Next, we investigated the effect of the sample chamber size on the resonance frequency of the PA sensor. The sample chamber sizes investigated were 8 mm × 3 mm, 6 mm × 3 mm, 4 mm × 3 mm, and 2 mm × 3 mm. The resonator column length in all cases was $R_L = 100$ mm. Figure 7(a) shows the measured resonance spectra normalized to the incident power.

The resonance frequency of the chamber was increased with decreasing sample chamber size. The increment was due to the combined effect of the sample chamber and resonator column. The combination acts as a single resonator column with a formation of the acoustic node at the microphone end and antinode at the sample chamber end. Figure 7(b) shows the numerical simulation results, and Table 4 shows the comparison of the numerical and experimental values.

The PA signal amplitude also increased with decreasing sample chamber size. The PA amplitude was 43.4 V/W larger in the case of 2 mm × 3 mm size compared to the 8 mm × 3 mm size. This is due to damping caused by the air column in the cell. The PA signal is inversely proportional to the cell volume. However, the cell of smaller size would create acoustic damping losses caused by cell boundaries, which would be observed as a broadening of the resonance peak.

Table 4 Comparison of resonance frequencies obtained from the experimental and numerical simulation for the various sample chamber sizes, using a fixed resonator length of 100 mm.

Sample chamber size	Resonant frequency (Hz) simulation	Resonant frequency (Hz) experiment	Percentage difference
8 mm × 3 mm	370	377	1.87
6 mm × 3 mm	480	453	5.78
4 mm × 3 mm	540	536	0.74
2 mm × 3 mm	610	613	0.49

The quality factor (Q_j) of the system was estimated from the broadening of the peak.

Table 5 shows the quality factor Q_j of the PA sensor for various chamber sizes which was obtained by fitting Eq. (6) to the resonance frequency response of the various PA sensors. The larger sample chamber exhibited a better-quality factor than the smaller ones. For example, the Sc of size 8 mm × 3 mm had a 37% greater quality factor than the 2 mm × 3 mm Sc.

We have investigated numerically the possible eigenmodes supported by the PA cell. The analytical expression 8 used to identify the type of eigenmodes excited. Table 6 reports the

Table 5 Quality factor of the PA sensor with various sample chamber sizes.

Sample chamber	Quality factor (Experiment)
With fixed $R_L = 100$ mm	
8 mm × 3 mm	2.52
6 mm × 3 mm	2.49
4 mm × 3 mm	2.49
2 mm × 3 mm	1.84

Table 6 Analytical results of eigenmodes supported by the resonator.

Resonator length (mm)	Mode 100 frequency (Hz)	Mode 010 frequency (Hz)	Mode 001 frequency (Hz)	Type
Sc (8 mm × 3 mm)				
100	1657	194.2×10^3	404×10^3	$\lambda/4$
200	829	194.2×10^3	404×10^3	$\lambda/4$
400	414	194.2×10^3	404×10^3	$\lambda/2$

Table 7 Comparison of experimental, numerical, and analytical results of eigenmodes supported by the resonator.

Resonator length (mm)	Resonant frequency (Hz) experiment	Resonant frequency (Hz) numerical	Resonant frequency (Hz) analytical
Sc (8 mm × 3 mm)			
100	377	370	414 Hz
200	231	240	207
400 mm	174 Hz	200 Hz	203

frequencies of the first three lower-order eigenmodes supported by the PA sensor. Only the mode (100) was able to observe in this case due to the modulation range exhibited by the chopper. The mode (100) represents the longitudinal acoustic mode. In our case, we observed the longitudinal mode of type $\lambda/4$ with the open end in the sample chamber and close end at the microphone end. Table 7 represents the comparison of experimental and analytical results. The difference compared to the analytical results was due to that in Eq. (5), we have not accounted for the resonance tube end correction in the expression.⁵⁴

Next, the resolution and molecular sensitivity of the system were calculated. The PA cell of dimensions ($S_c = 8 \text{ mm} \times 3 \text{ mm}$ and $R_l = 100 \text{ mm}$) was chosen for the study. A USAF 1951 resolution chart and RBC were used for resolution and molecular sensitivity measurements.

Figures 8(b) and 8(c) show the PA image of the group 7 element 6 and the plot of line spread function (LSF) of one of the components of the element. The lateral resolution of a system defined as its ability to distinguish the two nearest-neighboring objects clearly. This can be estimated experimentally by quantifying the edge blurring or edge spread function of the image. The region in the solid rectangular box in the PA image used to calculate the edge spread function (ESF). The EDF is a step function converted to a Gaussian profile to find the edge spread with the help of Gaussian fit. The EDF converted to Gaussian profile by differentiating the data. The differentiated EDF is called the LSF. The star marker in the figure represents the LSF. The Gaussian fit (solid red line) on the LSF gives a lateral resolution of $1.9 \mu\text{m}$.

The molecular-sensitivity measurement of the system is also called as noise equivalent sensitivity because it depends on the total noise of the system. It provides information about the minimum number of molecules required to generate a PA signal greater than the system noise. RBCs were used to measure the molecular sensitivity of the system because the RBC is the

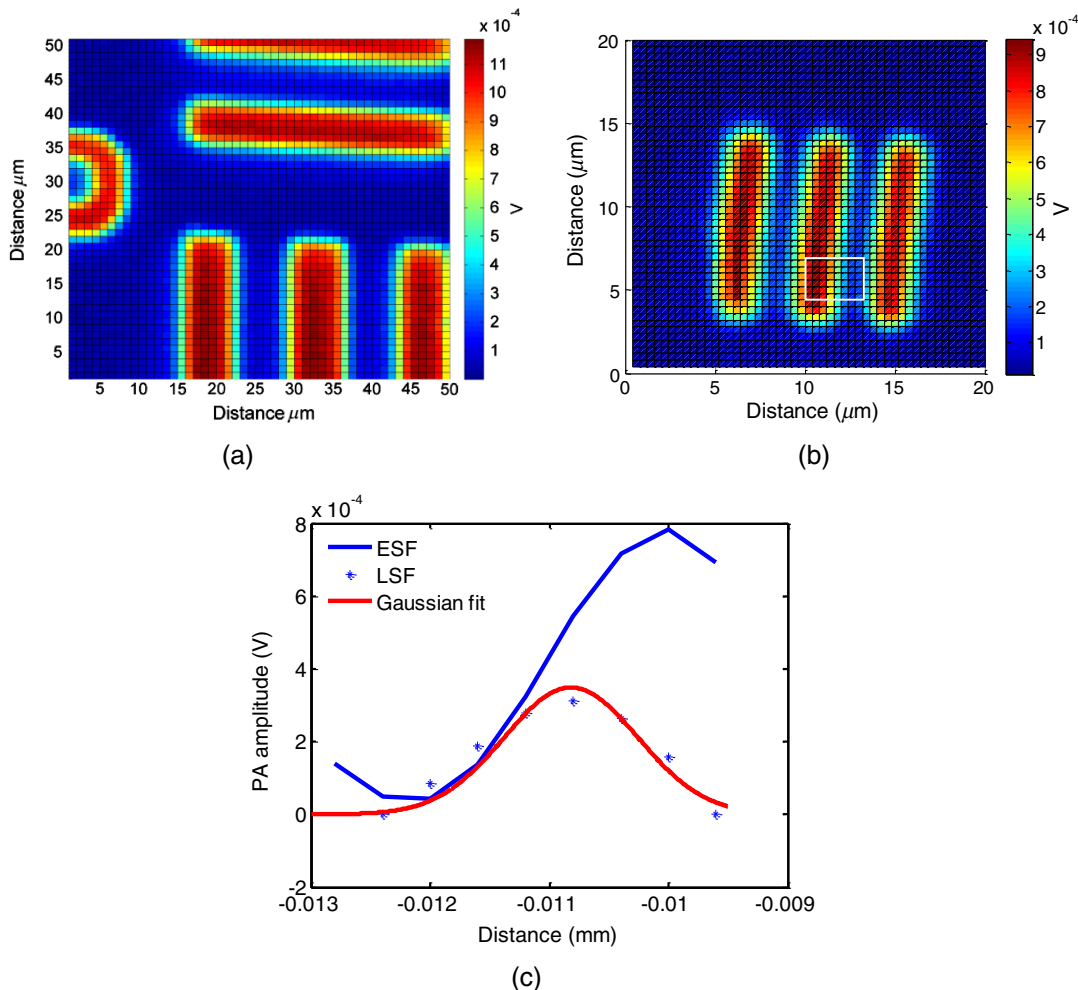


Fig. 8 PA image of a lateral resolution chart, USAF 1951 (a) group 6 element 6, (b) group 7 element 6, and (c) LSF, where the solid blue line is the ESF, marker * is the LSF, and the solid red line is Gaussian fit.

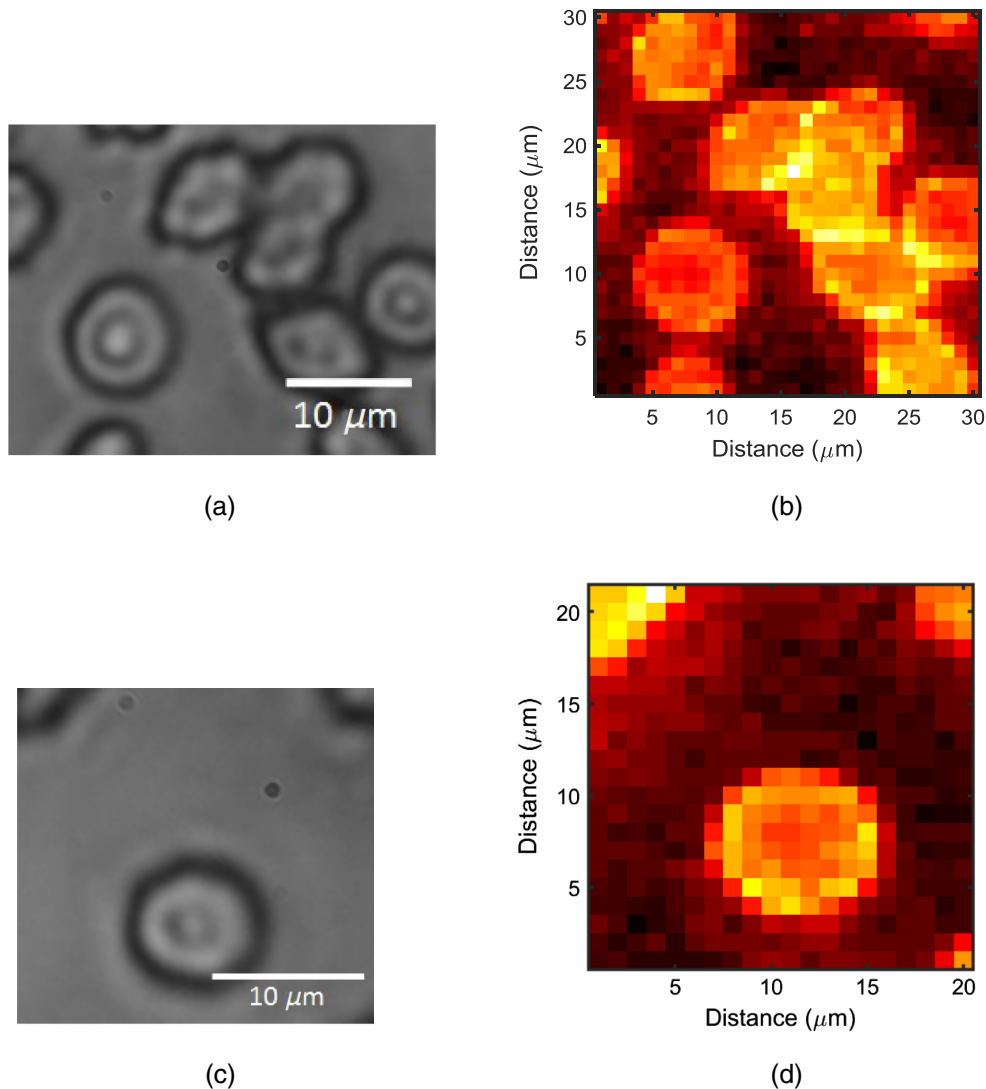


Fig. 9 (a) Optical and (b) PA image of blood smear on a microscope glass slide measured by trans-illumination method. (c) Optical and (d) PA image measured by the reflection method.

dominant endogenous chromophore in most PA imaging as it exhibits a very strong absorption cross-section in the optical and near-infrared wavelength.

Measurements performed on a common blood smear on a glass slide to probe single RBC. Figure 9 shows the optical and PA images of the RBCs. The measurement was also made in reflection mode [Fig. 9(b)] as conditions are relevant to *in vivo* measurements. Figure 9(d) shows the PA image measured in the reflection mode.

5 Discussion

Table 3 compares the numerical simulation and experiment results of PA resonance of various resonator lengths. The numerical simulation and experiment results show good agreement. The small difference in peak value between the experimental and simulation results could be due to the variation in mesh spacing between PA sensor of various sizes. The variation is bigger in the case of $RI = 400$ mm probably due to acoustic attenuation losses increase with traveling distance.

The amplitude of the resonance peak (Fig. 6) decreased with increasing in resonator length. The value decreased from

104.8 V/W for the resonator of length 400 mm to 43.86 V/W for the resonator of length 100 mm. The decrease in the PA amplitude could be due to attenuation of the acoustic signal by the thermo-viscous damping along the long resonator column.

We have investigated the attenuation loss due to PA cell chamber size. The PA cell of size 8 mm \times 3 mm (Table 5) exhibited a better quality factor compared to the smaller size. Thermal loss and acoustic damping can play a significant role in a smaller PA sample chamber. For example, air at room temperature and pressure has a thermal diffusion length of about 2 mm for a modulation frequency of 100 Hz, whereas thermo-viscous damping is 0.1 mm. Therefore, the sample chamber below the size of 2 mm should be avoided to study a material that has strong optical absorption and emissivity because PA signal saturation could be observed. However, a smaller PA chamber could be preferred to study samples of weaker optical absorption as it provides a stronger PA signal.

The PA cell also suffers from other losses such as volume loss and gas imperfection loss. The volume loss is caused by the friction of neighboring gas layers due to thermal conduction and viscosity in the gas volume.^{48,49}

Table 8 A summary of the sensor performance parameters.

Sensor property	Old sensor	New sensor
NEM	693×10^2	316×10^2
NEM per unit bandwidth	43,887 Hz ^{-1/2} or 72.8 zeptomoles	19,966 Hz ^{-1/2} or 33 zeptomoles
MPDL	19.1 μ Pa	8.7 μ Pa
MPDL per unit bandwidth	7.64 μ Pa/Hz ^{1/2}	3.5 μ Pa/Hz ^{1/2}

Finally, the resolution and the molecular sensitivity of the system were investigated. A USAF chart and RBCs were used for these purposes. The lateral resolution of the system was 1.9 μ m. The SNR of the system calculated by estimating the average PA signal with and without the RBC (Fig. 9). The point of interrogation was 1.2 μ m as a 40 \times objective was used for imaging. The average PA signal from the RBC and the background were 2.12×10^{-4} V and 9.54×10^{-7} V, respectively. The SNR of the system was ~ 220 .

Based on these measurements, the molecular sensitivity/molecular concentration equivalent to SNR was calculated. A single RBC contains about 270×10^6 hemoglobin molecules.⁵⁵ The mean corpuscular volume is 92.8×10^{-15} L. The estimated concentration of hemoglobin in a single RBC is $3.1 \times 10^6 \mu\text{m}^{-3}$. A 40 \times objective of 0.6 NA with an input light beam of a diameter of 0.36 mm excites about $\sim 7 \times 10^6$ hemoglobin molecules. The average PA signal detected was 2.12×10^{-4} V. The measured system molecular-sensitivity equivalent to noise [or noise equivalent sensitivity (NEM)] was $\sim 316 \times 10^2$. As the bandwidth of the lock-in amplifier filter was 1.5 Hz, the NEM per unit bandwidth (NEM/ $\sqrt{\Delta f}$) was $\sim 19,966 \text{ Hz}^{-1/2}$ or 33×10^{-21} mol or 33 zeptomoles. Table 8 summarizes the sensitivity and minimum detectable pressure of the system. The system exhibited sensitivity about 2.2 times more than our previously reported system.³⁸

Finally, the minimum pressure measurable by the system was calculated. Our microphone sensitivity was 50 mV/Pa. The PA signal was measured with zero amplifier gain. Therefore, the measured PA signal of the sensor is proportional to the pressure, i.e., 50 mV/Pa. The minimum measurable pressure is the pressure equivalent to NEM sensitivity of the system. The MPDL of the PA sensor was 8.7 μ Pa. The MPDL per unit bandwidth, $\text{MPDL}/\sqrt{\Delta f} = 3.4 \mu\text{Pa}/\text{Hz}^{1/2}$.

6 Conclusions

This study presents numerical simulations and experimental validation of a recently developed noncontact PA sensor for low-power CW laser-based frequency-domain PA microscopy. The main components of the PA sensor are the sample chamber and resonator column, which were optimized to maximize the signal. Various sample chamber and resonator column sizes were investigated. It was found that the resonance frequency of the PA cell decreased with an increase in the resonator length and that the resonance frequency increased with a decrease in the sample chamber size. The 8 mm \times 3 mm sample chamber exhibited a better quality factor compared to the other sizes examined (about 27% higher than the 2 mm \times 3 mm Sc). We attribute the drop in quality factor for the smaller cell to viscous-thermal

damping at the cell boundaries. Therefore, an Sc of minimum size 2 mm is suitable for PA imaging. The system exhibited NEM sensitivity per unit frequency was $\sim 19,966 \text{ Hz}^{-1/2}$ or 33×10^{-21} mol or 33 zeptomol and an MPDL per unit bandwidth, $\text{MPDL}/\sqrt{\Delta f} = 3.4 \mu\text{Pa}/\text{Hz}^{1/2}$. The developed PA sensor could be an inexpensive add-on component to existing commercial microscopes, which could use low-power CW or even pulse xenon lamp or LED sources. The developed PA sensor is cost effective and expected to have many applications, such as cellular imaging, surface tissue imaging, and sensing, and interrogating optically opaque samples and highly scattering sample such as powders.

Disclosures

The authors have no relevant financial interests in this article and no potential conflicts of interest to disclose.

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